Transient study of self-heating effects in AlGaIn/GaN HFETs: Consequence of carrier velocities, temperature, and device performance

Yuh-Renn Wu^{a)}Graduate Institute of Electro-Optical Engineering and Department of Electrical Engineering,
National Taiwan University, Taipei, Taiwan 10617, Taiwan

Jaspri Singh

Department of Electrical Engineering and Computer Science, University of Michigan,
Ann Arbor, Michigan 48109

(Received 25 September 2006; accepted 24 April 2007; published online 14 June 2007)

GaN transistors offer a unique combination of high speed and high power. Devices capable of operating at close to 180 GHz can also be biased so that each electron may emit up to 100 phonons as it transits from source to drain. Therefore, strong heating effects in GaN transistors are expected. In this article, we present a time dependent model to examine current, power and temperature in pulsed and direct current conditions. The studies show that self-heating effects are significant even for very short pulse widths ($\sim$ a few nanoseconds). The results explain part of the reasons for the low velocity observed in the short pulse measurements. Our calculations also suggest that the device performance could improve more than 30% if we can efficiently remove the self-heating effects.

© 2007 American Institute of Physics. [DOI: 10.1063/1.2745286]

I. INTRODUCTION

Over the last several years AlGaIn-GaN heterostructure field effect transistors (HFETs) have emerged as a very promising technology for microwave high frequency and high power applications.^{1–6} AlGaIn/GaN HFETs are high power devices where power dissipation reaching more than 10 W/mm (Ref. 6) is not uncommon. Therefore, strong self-heating effects in the AlGaIn/GaN channel are expected.^{7–10} The temperature increases in the channel could be as high as 200–300 K, so that the self-heating effect in the channel is a very important issue and cannot be ignored when the device is biased for large current flow.

Important issues that need to be understood are: What impact does heating have on current-voltage (I – V) characteristics, electron velocity, and on device performance? Are the self-heating effects adequately described by equilibrium phonons or are nonequilibrium phonons (hot phonons) important? Can we do short pulse measurements to eliminate self-heating and how short should the pulses be to accomplish this? Experimental measurements, when compared with theoretical study, suggest that self-heating effects are important even for pulses short as a few nanoseconds.¹¹ It has been suggested that hot phonons or nonequilibrium phonons may play a role in current (electron velocity) suppression even for short pulses.

However, recently, Bartker *et al.*¹² have shown that hot phonon effects may not be very serious and the suggested hot phonon life time is less than 0.3 ps. In their work, the 10 ns short pulse measurement is used and the decrease of velocity in the high field region is not as serious as previously reported results. Therefore, the effects of hot phonons are still unclear. As mentioned earlier, the hot phonon lifetime calcu-

lations are based on the assumption that the self-heating effects are completely removed in very short pulse measurement. However, due to technology limitation, a very short pulse measurement (<1 ns) is difficult. An important question is: When is the pulse short enough to completely remove self-heating effects?

In order to address these issues raised earlier, we have developed a two-dimensional (2D) time dependent thermal conduction equation coupled with our 2D Poisson and drift-diffusion equation solver¹³ to study the temperature evolution in time domain for the nitride HFET. The relation of current, power, and temperature with time evolution will be addressed in this article.

II. FORMALISM

In order to study self-heating effect especially for the steady state and transient state studies, we need to include the 2D thermal conduction equation into our developed 2D Poisson and drift-diffusion solver.¹³ The thermal conduction equations can be written as

$$C \frac{\partial T}{\partial t} = H + \nabla \cdot (\kappa \nabla T), \quad (1)$$

where κ is the thermal conductivity ($\text{W cm}^{-1} \text{K}^{-1}$), H is the heat generation term (W cm^{-3}), C is the heat capacity (J K cm^{-3}), and T is the temperature. The spatial domain of the thermal conduction equation is solved by the two-dimensional finite element method and the time domain is discretized by the central difference method, which has better accuracy and stability. For calculating the steady state solutions of I – V curves with the self-heating effect, the self-consistent iteration loops of thermal conduction, Poisson and drift-diffusion equations are used in order to obtain the convergence. The average temperature in the channel obtained from the thermal conduction equation and temperature de-

^{a)}Electronic mail: yrwu@cc.ee.ntu.edu.tw

TABLE I. Material parameters used to calculate the thermal conduction equation.

	Thermal conductivity	Heat capacity
GaN	1.3 W/(cm K)	2.97 J/cm ³ K
SiC	3.3 W/(cm K)	2.10 J/cm ³ K
Sapphire	0.35 W/(cm K)	1.66 J/cm ³ K

pendent velocity-field (v - E) characteristics are used in the Poisson and drift-diffusion solver in order to maintain the program stability. To study the time dependent problem, the Poisson and drift-diffusion equations are solved without heating effects in the first step. Then the time dependent thermal conduction equation is solved at time Δt . The temperature increase in the channel is fed back to the Poisson and drift-diffusion solver again to obtain the current changes in the device. The coupled equations are solved for each time step up to the intended time.

III. SIMULATION RESULTS

The device structures of interest are the traditional $\text{Al}_{0.32}\text{Ga}_{0.68}\text{N}/\text{GaN}$ HFETs and the two terminal $\text{Al}_{0.32}\text{Ga}_{0.68}\text{N}/\text{GaN}$ HFETs without the gate, which are used to examine v - E characteristics experimentally. The channel length is $1.6\text{ }\mu\text{m}$. The GaN buffer layer is $3\text{ }\mu\text{m}$ and the SiC substrate is approximately $100\text{ }\mu\text{m}$ thick unless otherwise stated. The $\text{Al}_{0.32}\text{Ga}_{0.68}\text{N}$ layer thickness is 25 nm and the passivation layer is not included in the simulation. The top of the device is assumed to be air and behaves as a thermal insulator. The v - E characteristics at different temperatures are calculated using the Monte Carlo method.¹⁴ The v - E characteristics as a function of temperature are then interpolated from these curves for different electric field. Finally, the temperature dependent v - E characteristics are used to calculate the current at different temperatures in the channel. The material parameters used for thermal conduction equations are also listed in Table I.

In order to develop confidence in the thermal equation solver, we calculate the thermal resistance versus substrate thickness and compare to experimental data. Figure 1 shows the thermal resistance for a SiC substrate with varying thickness. As shown in the Fig. 1, the thermal resistance decreases as the SiC substrate thickness decreases. When the GaN buffer layer becomes thinner, the thermal resistance also becomes lower due to the lower thermal conductivity of GaN compared to SiC. The typical experimental values of thermal resistance vary from 5.0 to 12.9 K mm/W , which are in the range of experimental measurements.^{8,15} Therefore, our results are in reasonable ranges. It should be noted that our simulation is based on 2D approximation, which is suitable for very wide channel cases. Therefore, slight differences from experimental results are expected but it should not affect explaining general trends. For the small channel width, the thermal resistance may drop 10% – 30% .

Before discussing the actual device operation under various bias conditions, it is important to develop a general idea of how the temperature increases in the channel for various

FIG. 1. The thermal resistance vs SiC substrate thickness for AlGaIn/GaN HFETs with different GaN buffer layer thickness values.

power density levels. To calculate the heating time constant, we only need to solve the time-dependent thermal conduction equation by assuming various constant channel power densities at an initial time. Figure 2 shows the predicted effects of (a) how channel temperature evolves with time for various constant power density levels and (b) what the channel temperature is at various times for different as a function of power density. Note that typical values of highest power densities in GaN HFETs are around 5 – 30 W/mm^2 .^{2–6} We see that significant temperature changes occur in the device even during very short times. As shown in Fig. 2(a), the temperature increases rapidly in a few nanoseconds and then increases gradually as time increases. Figure 2(b) shows the average channel temperature versus power density for differ-

FIG. 2. (a) The simulated channel temperature increase vs time for different output power densities. After 200 ns , the heating source is removed and the temperature decreases rapidly and (b) the simulated average channel temperature vs power density for different pulse width values.

FIG. 3. The temperature distribution along the z direction of the two terminal ungated device for different time periods. The GaN buffer layer is $3 \mu\text{m}$ and SiC substrate is $100 \mu\text{m}$.

ent short pulses. It is shown that temperature change ΔT for 200 ns is approximately 60% of the steady state value so that the heating problem for 200 ns is still very serious and cannot be ignored. The calculations also suggest that the heating time constants are in a few nanoseconds range in the high power density conditions. The average channel temperature is predicted to increase 20–70 K even for a 4 ns pulse, for a variety of power density values. Here, we only calculate the temperature increase with constant power density. In actual devices, the rate at which the temperature increases will be affected by the changes of current due to heating and biasing conditions and we will discuss it later.

It is important to know how fast heat propagates to the buffer region. If the heat does not propagate across the GaN/substrate interface in the short pulse period region, it should be concluded that the substrate does not play an important role in removing heat in the short pulse measurement. Therefore, improved designs to assist heat dissipation may be needed in these cases. Figure 3 shows the predicted temperature distributions in the device for different time steps at $V_D = 10 \text{ V}$. The results are calculated by solving the Poisson and drift-diffusion equation with a time dependent thermal conduction equation self-consistently. The device structure is an ungated two terminal device as shown in Fig. 3. The GaN buffer layer thickness is $3 \mu\text{m}$. As shown in Fig. 3, at $t = 10 \text{ ns}$, the heat generated in the channel has not had a chance to propagate across the GaN/SiC interface while at $t = 100 \text{ ns}$, the heat just started to pass through the GaN/SiC interface. The differences of temperature distribution for transient and steady state ($t = \text{infinity}$) are significant, especially in the substrate region. This suggests that for the short pulse periods ($< 100 \text{ ns}$), the substrate does not play any role in the removal of heat. Therefore, for such short pulses, the experimental results for either sapphire or the SiC substrate are expected to show the same self-heating effects. This is indeed seen in Ref. 12 with the sapphire substrate and 10 ns pulse measurement.

We now address the $v-E$ relationship. It is common to use the $I-V$ curves in an ungated two terminal device. Using a variety of V_{DS} values, we can calculate the current and electric field in the channel. The carrier velocity is then ex-

FIG. 4. The current density and average channel temperature vs time for different drain bias values. The line with symbol represents the temperature and the line without symbol represents the current.

tracted from the current. To study the impact of self-heating, the time dependent thermal conduction solver coupled with the Poisson and drift-diffusion solver is used. Figure 4 shows the current density and temperature versus time for various V_{DS} values. When the drain voltage increases, we observe a rapid decrease of current at $t < 5 \text{ ns}$. The temperature in the channel increases 150 K and the current drops 25% at $t = 5 \text{ ns}$ for V_{DS} is equal to 20 V. The rate of decrease of the current for time greater than 100 ns is much smaller compared to the current drop within 5 ns. Thus, these simulations predicted that if we use short pulses ranging from 100 ns to $1 \mu\text{s}$ to measure the $v-E$ curves, the self-heating effects are not completely eliminated even though the current does not change too much within this region.

Figure 5 shows $v-E$ characteristics predicted for different pulse lengths. The velocity has a significant drop especially in the high field region even for a 3 ns pulse. For the pulse lengths around 200 ns, which is the typical pulse length used for $I-V$ curves measurement of an AlGaIn/GaN HFET, the predicted saturation velocity reaches $1.50 \times 10^7 \text{ cm/s}$. This value is very close to the velocity extracted from the extrinsic unit gain cut-off frequency f_T measurement⁴ and is

FIG. 5. The simulated $v-E$ curves for different short pulse cases. The simulation shows that even with a 3 ns pulse width, the self-heating effect cannot be completely removed especially for high field regions.

FIG. 6. (a) Comparison showing the consequence of self-heating on calculated g_m curves at $V_{DS}=10$ V for SiC substrate, and (b) the comparison of calculated g_m curves at $V_{DS}=10$ V for sapphire substrates. The results of the 200 ns pulse are included in the simulation.

probably one of the reasons for lower current values. However, our simulation results still show a slightly higher predicted velocity compared with the experimental work published in Ref. 11. Therefore, the effects of hot phonons are not fully excluded if there is no evidence of other factors limiting the performance such as alloy scattering due to hot carriers, etc. But our results suggest that to fit the hot phonon life, one needs to consider the actual channel temperature instead of the ambient temperature.

Figure 6 shows the predicted transconductance g_m for AlGaIn/GaN HFETs including a 200 ns pulse with 0.25 μm gate length. As shown in Fig. 6, for SiC substrates, the difference of g_m between a 200 ns pulse and the dc is very small, while the difference for the sapphire substrates is very significant. These are also observed experimentally.¹⁶ The predicted 200 ns pulse g_m for the SiC substrates only shows a slight difference from the sapphire substrates. The simulations suggest that for a 200 ns pulse, the heat dissipation is mainly affected by the GaN buffer layer thickness. Therefore, if we can reduce the GaN buffer thickness, we should observe more differences for a 200 ns pulse measurement⁷ with different substrates.

IV. DISCUSSION

The use of the SiC substrate has improved the device performance significantly compared to the sapphire substrate. However, our results suggest that even for short pulse measurement in the SiC substrate, the current and electron velocity are still suppressed by 10%–30% to the ideal case (without heating effects). As we know, the device performances such as g_m , f_T , sheet resistance, and out put power density are closely related to the electron velocity. Although f_T up to 180 GHz has been reported,¹⁷ the operation voltage is small ($V_{DS}=5$ V) to avoid the self-heating effect and the drain delay.¹⁴ Therefore, the device output performance is seriously limited by the self-heating effect even at short pulse measurements, especially at high frequency and high power conditions. And the possible device performance improvement could be more than 30% if the heating effect is removed.

There are several ways to suppress self-heating effects. As shown in Fig. 2, to suppress the self-heating, not only high thermal conductive materials such as SiC are needed, but the high thermal conductive materials also need to be as close to the GaN channel as possible. One is to use a thinner GaN buffer as well as a thinner SiC substrate as shown in Fig. 1. The benefits of this approach have also been observed by experiments,⁷ where a 0.5 μm GaN buffer is used to reduce the thermal resistance to 2 K mm/W. However, a thinner GaN buffer layer may lead to higher dislocations due to the lattice mismatch to the SiC substrate. The other possible way is to apply the wafer bonding technologies. Once the nitride HFETs are made, we can remove the substrate and bond the film to high thermal conductivity thin film such as diamond. This will greatly help to reduce the self-heating effect. Depositing a higher thermal conductive layer on top of nitride HFETs will also help to reduce the self-heating. For extraction of the hot-phonon lifetime, the channel temperature rather than the ambient temperature needs to be considered.

It is important to note that if the power density of the GaN device is high enough, the optical phonon emission can become so large that the phonon number can buildup beyond the value given by equilibrium conditions. This is especially true if the optical phonon lifetime is long. Since the polar optical phonon scattering rates are much higher than acoustic phonon scattering rates, the higher population of the optical phonon density will result in a lower carrier velocity. Therefore, the determination of the hot-phonon lifetimes becomes an important issue. The phonon lifetime might also be affected by the crystal quality of the GaN device and may be shorter for imperfect crystal structures. Therefore, experimental results carried out on devices where the crystal qualities are different should not be compared assuming the same phonon lifetime. Our simulations also suggest that self-heating effects should not be ignored in estimating the hot-phonon lifetime. With the proper choice of short pulse experiments, it should be possible to suppress the hot-phonon effect.

V. CONCLUSION

Our simulations suggest that self-heating cannot be completely eliminated even at very short pulse lengths. Improved device engineering is needed to reduce the self-heating effect. A comparison of experiments and simulation reveals higher predicted velocity in the presence of self-heating effect. In the absence of significant alloy scattering, this suggests a substantial role of hot phonons. For extraction of the hot-phonon lifetime, the channel temperature rather than the ambient temperature needs to be considered.

ACKNOWLEDGMENTS

The authors would like to thank Dr. Darwish for very useful discussions. This work was supported by the U. S. Office of Naval Research MURI program (Grant Nos. F013465 and F004815).

¹M. Higashiwaki, T. Matsui, and T. Mimura, IEEE Electron Device Lett. **27**, 16 (2006).

²M. Kanamura, T. Kikkawa, T. Iwai, K. Imanishi, T. Kubo, and K. Joshin, Tech. Dig.-Int. Electron Devices Meet. **2005**, 572-575 (2005).

³U. K. Mishra, P. Parikh, and Y. F. Wu, Proc. IEEE **90**, 1022 (2002).

⁴T. Palacios, E. Snow, Y. Pei, A. Chakraborty, S. Keller, S. P. Denbaars,

and U. K. Mishra, Tech. Dig.-Int. Electron Devices Meet. **2005** (2005).

⁵R. Therrien *et al.*, Tech. Dig.-Int. Electron Devices Meet. **2005**, 568-571 (2005).

⁶Y. F. Wu, M. Moore, A. Saxler, T. Wisleder, U. K. Mishra, and P. Parikh, Tech. Dig.-Int. Electron Devices Meet. **2005**, 583-585 (2005).

⁷R. Gaska, A. Osinsky, J. Yang, and M. Shur, IEEE Electron Device Lett. **19**, 89 (1998).

⁸M. Kuball, J. M. Hayes, M. J. Uren, I. Martin, J. C. H. Birbeck, R. S. Balmer, and B. T. Hughes, IEEE Electron Device Lett. **23**, 7 (2002).

⁹T. J. Anderson, F. Ren, L. Covert, J. Lin, and S. J. Pearton, J. Vac. Sci. Technol. B **24**, 284 (2006).

¹⁰A. M. Darwish, A. J. Bayba, and H. A. Hung, IEEE Trans. Microw. Theory Tech. **52**, 2611 (2004).

¹¹L. Ardaravičius, A. Matulionis, J. Liberis, O. Kiprijanovic, M. Ramonas, L. F. Eastman, J. R. Shealy, and A. Vertiatchikh, Appl. Phys. Lett. **83**, 4038 (2003).

¹²J. M. Barker, D. K. Ferry, D. D. Koleske, and R. J. Shul, J. Appl. Phys. **97**, 063705 (2005).

¹³Y.-R. Wu, M. Singh, and J. Singh, IEEE Trans. Electron. Devices **52**, 1048 (2005).

¹⁴Y. R. Wu, M. Singh, and J. Singh, IEEE Trans. Electron. Devices **53**, 588 (2006).

¹⁵M. Kuball, S. Rajasingam, A. Sarua, M. J. Uren, T. Martin, B. T. Hughes, K. P. Hilton, and R. S. Balmer, Appl. Phys. Lett. **82**, 124 (2003).

¹⁶L. Shen, I. P. Smochkova, D. S. Green, S. Heikman, and U. K. Mishra, J. Vac. Sci. Technol. B **21**, 540 (2003).

¹⁷M. Higashiwaki, T. Matsui, and T. Mimura, 2006 Device Research Conference, 2006.